CONTROLLABILITY CRITERIA FOR TIME-DELAY FRACTIONAL SYSTEMS WITH A RETARDED STATE

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The paper is concerned with time-delay linear fractional systems with multiple delays in the state. A formula for the solution of the discussed systems is presented and derived using the Laplace transform. Definitions of relative controllability with and without constraints for linear fractional systems with delays in the state are formulated. Relative controllability, both with and without constraints imposed on control values, is discussed. Various types of necessary and sufficient conditions for relative controllability and relative $U$-controllability are established and proved. Numerical examples illustrate the obtained theoretical results.

Keywords: fractional dynamical systems, controllability, delays in the state, constraints, pseudo-transition matrix, Caputo derivative.

1. Introduction

Fractional-order derivatives are a generalization of classical integer-order ones. Mathematical modeling of systems and processes with the use of fractional-order derivatives leads to fractional differential equations. Fractional differential equations occur in mathematical models of, among other things, mechanical, biological, chemical and medical phenomena. It has become apparent that fractional-order models reflect the behavior of many real-life processes more accurately than integer-order ones. For more details concerning fractional calculus and its practical applications refer to the monographs of Oldham and Spanier (1974), Miller and Ross (1993), Samko et al. (1993), Podlubny (1999), Kilbas et al. (2006), Sabatier et al. (2007) or Monje et al. (2010).

Numerous mathematical models describe dynamical systems with delays in control, or both the state and control. Therefore, studying the properties of systems with delays is especially important.

The controllability of dynamical systems plays a crucial role in their analysis. In recent years, various controllability problems for different types of fractional-order dynamical systems have been considered in many publications and monographs. The controllability of deterministic fractional dynamical systems without delays was studied, among others, by Klamka (2011), Klamka et al. (2014) or Babiarz et al. (2016) for discrete-time fractional systems, Kaczorek (2011) as well as Kaczorek and Rogowski (2015) for positive fractional linear systems, both discrete- and continuous-time, and Chen et al. (2006), Chikriy and Matichin (2008), Sakthivel et al. (2011), Wang and Zhou (2012), Balachandran and Kokila (2012; 2013) or Balachandran et al. (2012b) for continuous time fractional systems.

The controllability of fractional systems with delays in control was analyzed by Trzasko (2008), Kaczorek (2011), Balachandran et al., (2012c; 2012a), Wei (2012) as well as Kaczorek and Rogowski (2015). The controllability of fractional systems with delays in the state was discussed by Zhang et al. (2013) and Busłowicz (2014). However, there are only few works concerning the controllability of time-delay fractional systems with retarded state. It should also be noted that the majority of papers on controllability of fractional systems address controllability issues for unconstrained controls. Constrained controllability of integer order systems with delays was discussed, among others, by Sikora (2003; 2005), Klamka (2008; 2009), or Sikora and Klamka (2012). Works on controllability of linear fractional systems with bounded inputs include those by Kaczorek (2014a; 2014b) for fractional positive discrete-time linear systems and fractional positive continuous-time...
linear systems, respectively, Sikora (2016) for fractional continuous-time linear systems, as well as Pawlusiewicz and Mozyrska (2013) for h-difference linear systems with two fractional orders. It should be stressed that, in practice, control (an input function) is not completely unlimited, but is usually constrained in various ways.

The aim of the paper is to give new controllability criteria (necessary and sufficient conditions) for continuous-time linear fractional systems with delays in the state. The controllability criteria both for unconstrained and constrained controls are formulated and proved. Theoretical results presented in the paper can be applied, among other things, to chemical solution control systems. For example, the cascade connection of two fully filled mixers can be described by a system of fractional equations with one delay in the state.

The paper is organized as follows. In Section 2 we recall some preliminary definitions and formulas. In Section 3 we present the mathematical model of linear fractional dynamical systems with multiple delays in the state. We formulate and prove the existence theorem for the solution of the discussed systems. Section 4 contains the main results of the paper. First, we formulate definitions for relative controllability and relative $U$-controllability of systems. Next, the main results of the paper are presented, i.e., the criteria (necessary and sufficient conditions) for relative controllability of the examined time-delay fractional systems with a retarded state. The proofs of the theorems are given. In Section 5 the theoretical results are illustrated by numerical examples. Finally, some concluding remarks are included in Section 6.

2. Preliminaries

Mathematical models containing fractional differential equations turn out to better describe some phenomena previously modeled by integer-order differential equations. Different fractional derivatives have been defined in fractional calculus. In this paper we use the Caputo fractional derivative. This is due to the fact that in the Caputo approach the initial conditions for fractional differential equations take on the same form as for integer-order differential equations (Podlubny, 1999). The definition is the following.

**Definition 1.** The Caputo fractional derivative of the order $\alpha$ ($n < \alpha < n + 1, n \in \mathbb{N}$) for a given function $f: \mathbb{R}^+ \to \mathbb{R}$ is the function

$$C^D^\alpha f(t) = \frac{1}{\Gamma(n - \alpha + 1)} \int_0^t \frac{f^{(n+1)}(\tau)}{(t-\tau)^{\alpha-n}} \, d\tau,$$

where $\Gamma$ is the gamma function.

It is obvious that as $\alpha \to n$ the Caputo derivative tends to the $n$-th order conventional derivative of the function $f$, e.g., $\lim_{\alpha \to n} C^D^\alpha f(t) = f^{(n)}(t)$.

As has been mentioned, the Caputo fractional derivative allows traditional initial conditions to be considered in the formulation of a mathematical model of a dynamical system.

In the theory of fractional calculus, an important role is that of the Mittag-Leffler function. Below we recall the definition (Podlubny, 1999).

**Definition 2.** The Mittag-Leffler function is a special function defined by the following series:

$$E_{\alpha,\beta}(z) = \sum_{k=0}^{\infty} \frac{z^k}{\Gamma(k\alpha + \beta)},$$

for $z \in \mathbb{C}$ and $\alpha, \beta > 0$.

For real positive $\alpha$ and $\beta$, the series in Definition 2 converges for all values of the argument $z$.

Based on the definition of the Mittag-Leffler function, for $\alpha > 0$ and an arbitrary $n$-th order square matrix $A$, we can give the formula for a pseudo-transition matrix $\Phi_0(t)$ of the linear fractional system $C^D^\alpha x(t) = A(t)x(t)$ (Monje et al., 2010):

$$\Phi_0(t) = E_{\alpha,1}(At^\alpha) = \sum_{k=0}^{\infty} \frac{A^k t^{\alpha k}}{\Gamma(k\alpha + 1)},$$

and then we set

$$\Phi(t) = t^{\alpha-1} E_{\alpha,\alpha}(At^\alpha) = t^{\alpha-1} \sum_{k=0}^{\infty} \frac{A^k t^{\alpha k}}{\Gamma((k+1)\alpha)}.$$

For $\alpha = 1$ we obtain the classical transition matrix of ordinary differential equations,

$$\Phi_0(t) = \sum_{k=0}^{\infty} \frac{A^k t^k}{\Gamma(k+1)} = \sum_{k=0}^{\infty} \frac{(At)^k}{k!} = e^{At}.$$

Therefore the pseudo-transition matrix $\Phi_0(t)$ is also called the matrix $\alpha$-exponential function and is denoted by $\Phi_0(t) = e^{At}_{\alpha}$ (Kilbas et al., 2006). It is convergent in the space of the $n$-th order matrices with real elements.

For the purposes of further calculations there are some formulas for the Laplace and the inverse Laplace transforms. For $\alpha \in (0, 1)$, the following formulas hold:

$$\mathcal{L}[C^D^\alpha f(t)] = s^\alpha \mathcal{L}[f(t)] - s^{\alpha-1} f(0)$$

(see, e.g., Zhang et al., 2013) and

$$\mathcal{L}^{-1}[s^{\alpha-1}(s^\alpha I - A)^{-1}] = \Phi_0(t),$$

$$\mathcal{L}^{-1}[(s^\alpha I - A)^{-1}] = \Phi(t)$$

(see, e.g., Monje et al., 2010).
3. Mathematical model

We consider time-delay linear fractional dynamical systems with multiple, constant delays in the state described by a fractional differential equation with a retarded argument of the following form:

\[ C^{D^{\alpha}}x(t) = \sum_{i=0}^{M} A_i x(t - h_i) + Bu(t) \]  \hspace{1cm} (1)

for \( t \geq 0 \) and \( 0 < \alpha < 1 \), where

- \( x(t) \in \mathbb{R}^n \) is the state vector,
- \( u \in L^2_{\text{loc}}([0, \infty), \mathbb{R}^m) \) is the control,
- \( A_i \) are \((n \times n)\)-dimensional matrices with real elements for \( i = 0, 1, \ldots, M \),
- \( B \) is an \((n \times m)\)-dimensional matrix with real elements,
- \( h_i : [0, T] \rightarrow \mathbb{R} \) for \( i = 1, 2, \ldots, M \) are constant delays in the state such that the following inequalities hold:
  \[ 0 = h_0 < h_1 < \ldots < h_i < \ldots < h_{M-1} < h_M. \]

Let \( z_0 = (x(0), x_0) \in \mathbb{R}^n \times L^2([-h_M, 0], \mathbb{R}^n) \) be the initial conditions, where \( x(\cdot) \in \mathbb{R}^n \) is the state vector at time \( t = 0 \) and \( x_0 \) is a function given on \([-h_M, 0]\). The Hilbert space \( \mathbb{R}^n \times L^2([-h_M, 0], \mathbb{R}^n) \) with the scalar product defined as

\[
\langle (x(t), y_1), (y(t), y_2) \rangle = \sum_{i=0}^{n} x_i(t)y_i(t) + \int_{-h_M}^{0} \langle x_i(\tau), y_i(\tau) \rangle_{\mathbb{R}^n} \, d\tau
\]

is denoted by \( M^2([-h_M, 0], \mathbb{R}^n) \).

The pair \( z_0 = (x(0), x_0) \in M^2([-h_M, 0], \mathbb{R}^n) \), where \( x(0) \) is the state vector at time \( t = 0 \) and \( x_0 = x_0(\tau) \) for \( \tau \in [-h_M, 0] \) is the segment of the trajectory of the length \( h_M \) defined on \([t - h_M, t]\), is called the initial complete state of the fractional system (1) for \( t \geq 0 \).

Any control \( u \in L^2_{\text{loc}}([0, \infty), \mathbb{R}^m) \) is called an admissible control for the fractional system (1).

**Theorem 1.** For given initial conditions \( z_0 = (x(0), x_0) \) and an admissible control \( u \), for every \( t \geq 0 \) there exists a unique solution \( x(t) = x(t, z_0, u) \) of the fractional equation (1) of the form

\[
x(t) = \Phi_0(t)x(0) + \int_{0}^{t} \Phi(t - \tau) \sum_{i=1}^{M} A_i x(\tau - h_i) \, d\tau + \int_{0}^{t} \Phi(t - \tau) Bu(\tau) \, d\tau,
\]

where

\[
\Phi_0(t) = \sum_{k=0}^{\infty} A_k t^{\alpha k} \frac{1}{\Gamma(k\alpha + 1)},
\]

\[
\Phi(t) = t^{\alpha - 1} \sum_{k=0}^{\infty} A_k t^{\alpha k} \frac{1}{\Gamma((k+1)\alpha)}.
\]

**Proof.** Applying the Laplace transform to both sides of the fractional differential equation (1), for \( t \geq 0 \) we have

\[
s^{\alpha} \mathcal{L}[x(t)] - s^{\alpha - 1} x(0)
\]

\[
= A_0 \mathcal{L}[x(t)] + \mathcal{L} \left[ \sum_{i=1}^{M} A_i x(t - h_i) + Bu(t) \right],
\]

Hence

\[
(s^{\alpha} I - A_0) \mathcal{L}[x(t)]
\]

\[
= s^{\alpha - 1} x(0) + \mathcal{L} \left[ \sum_{i=1}^{M} A_i x(t - h_i) + Bu(t) \right],
\]

and thus

\[
\mathcal{L}[x(t)] = (s^{\alpha} I - A_0)^{-1} s^{\alpha - 1} x(0)
\]

\[
+ (s^{\alpha} I - A_0)^{-1} \mathcal{L} \left[ \sum_{i=1}^{M} A_i x(t - h_i) + Bu(t) \right]
\]

\[
= \mathcal{L}[\Phi_0(t)x(0)] + \mathcal{L}[\Phi(t)] \mathcal{L} \left[ \sum_{i=1}^{M} A_i x(t - h_i) + Bu(t) \right].
\]

Now we apply the convolution theorem for the Laplace transform and obtain

\[
\mathcal{L}[x(t)] = \mathcal{L}[\Phi_0(t)x(0)]
\]

\[
+ \mathcal{L} \left[ \int_{0}^{t} \Phi(t - \tau) \left( \sum_{i=1}^{M} A_i x(\tau - h_i) + Bu(\tau) \right) \, d\tau \right].
\]

Using the inverse Laplace transform on both the sides of the above equation, we have

\[
x(t) = \Phi_0(t)x(0) + \int_{0}^{t} \Phi(t - \tau) \left( \sum_{i=1}^{M} A_i x(\tau - h_i) + Bu(\tau) \right) \, d\tau
\]

\[
= \Phi_0(t)x(0) + \int_{0}^{t} \Phi(t - \tau) \sum_{i=1}^{M} A_i x(\tau - h_i) \, d\tau + \int_{0}^{t} \Phi(t - \tau) Bu(\tau) \, d\tau.
\]

\[ \square \]
4. Main results

In this section we formulate and prove some theorems that establish the criteria (necessary and sufficient conditions) of relative controllability and relative \( U \)-controllability for the time-delay fractional system (1).

Before we formulate some controllability criteria for the time-delay fractional system (1), in this section we define relative controllability and relative \( U \)-controllability for the system (1) on the time interval \([0, T]\).

**Definition 3.** The time-delay fractional system (1) is called **complete state** if, for every \( \tilde{x} \in S \), there exists an admissible control \( \tilde{u} \in L^2([0, T], \mathbb{R}^m) \) such that

\[
x(T, z_0, \tilde{u}) = \tilde{x}.
\]

**Definition 4.** The time-delay fractional system (1) is called **relatively controllable** on \([0, T]\) from the initial complete state \( z_0 = (x(0), x_0) \) into a set \( S \subset \mathbb{R}^n \) if, for every \( \tilde{x} \in S \), there exists an admissible control \( \tilde{u} \in L^2([0, T], U) \), \( U \subset \mathbb{R}^m \), such that

\[
x(T, z_0, \tilde{u}) = \tilde{x}.
\]

If \( S = \{0\} \), then the system is called relatively null controllable or relatively null \( U \)-controllable. If \( S = \mathbb{R}^n \), the system is called relatively controllable or relatively \( U \)-controllable, respectively.

The theorem below formulates the necessary and sufficient conditions for relative controllability of the retarded system (1).

**Theorem 2.** The fractional system (1) is relatively controllable on the time interval \([0, T]\) if and only if the \((n \times n)\)-dimensional Gramian matrix

\[
W(0, T) = \int_0^T \Phi(T - \tau)BB^t\Phi'(T - \tau) d\tau
\]

is nonsingular, where \( \Phi' \) denotes the matrix transpose and \( \Phi(t) = E_{\alpha, \alpha}(A_0^t) \).

**Proof.** We prove the sufficient condition first. Suppose that \( W(0, T) \) is nonsingular. It follows that there exists the inverse matrix \( W^{-1}(0, T) \). For any initial complete state \( z_0 \in \mathbb{R}^n \times L^2([-h_M, 0], \mathbb{R}^n) \) we can take the following control function:

\[
\tilde{u}(t) = B^t\Phi'(T - t)W^{-1}(0, T)\left[\tilde{x} - \Phi_0(t)x(0)\right] - \int_0^T \Phi(T - s)\sum_{i=1}^M A_i x(s - h_i) \, ds
\]

for \( \tilde{x} \in \mathbb{R}^n \). From Theorem it follows that

\[
x(T) = x(T, z_0, \tilde{u}) = \Phi_0(T)x(0) + \int_0^T \Phi(T - \tau) \sum_{i=1}^M A_i x(\tau - h_i) \, d\tau + \int_0^T \Phi(T - \tau)B\tilde{u}(\tau) \, d\tau.
\]

After substitution we obtain

\[
x(T, z_0, \tilde{u}) = \Phi_0(T)x(0) + \int_0^T \Phi(T - \tau) \sum_{i=1}^M A_i x(\tau - h_i) \, d\tau + \int_0^T \Phi(T - \tau)B\tilde{u}(\tau) \, d\tau.
\]

By Definition the system (1) is relatively controllable on \([0, T]\).

We prove the necessary condition by contradiction. Suppose that the system (1) is relatively controllable, but the Gramian matrix \( W(0, T) \) is singular. Then there exists a vector \( \tilde{x} \neq 0 \) such that

\[
\tilde{x}'W(0, T)\tilde{x} = 0 = \int_0^T \tilde{x}'\Phi(T - t)BB^t\Phi'(T - t)\tilde{x} \, d\tau.
\]

Thus, for \( t \in [0, T] \), we have

\[
\tilde{x}'\Phi(T - t)B = 0. \tag{3}
\]

Since the system is controllable, it can be driven from the initial state \( z_0 \) to an arbitrary state \( x(T) \in \mathbb{R}^n \). Hence there exists a control \( u_0(t) \) that drives the initial state \( z_0 \) to zero. This means that

\[
x(T) = x(T, z_0, u_0) = \Phi_0(T)x(0) + \int_0^T \Phi(T - \tau) \sum_{i=1}^M A_i x(\tau - h_i) \, d\tau + \int_0^T \Phi(T - \tau)Bu_0(\tau) \, d\tau = 0.
\]

Moreover, there exists a control \( \tilde{u}(t) \) that drives the initial
Consider the linear mapping

\[
\tilde{x} = x(T, z_0, \tilde{u}) = \Phi_0(T)x(0) + \int_0^T \Phi(T - \tau) \sum_{i=1}^{M} A_i x(\tau - h_i) d\tau + \int_0^T \Phi(T - \tau) B \tilde{u}(\tau) d\tau.
\]

Combining the above two solutions, we have

\[
\tilde{x} - \int_0^T \Phi(T - \tau) B [\tilde{u}(\tau) - u_0(\tau)] d\tau = 0.
\]

Multiplying both the sides of the equality by \(\tilde{x}'\), and using (3), it follows that \(\tilde{x}'\tilde{x} = 0\). Thus \(\tilde{x} = 0\), which contradicts the assumption. Therefore the Gramian matrix \(W(0, T)\) is nonsingular.

Now we rewrite the solution (2) in the form containing the so-called free solution of the system (1). Let \(\Phi_f(t)\) be defined as

\[
\Phi_f(t) = \frac{t^{-\alpha}}{\Gamma(1 - \alpha)} I + \sum_{i=1}^{M} \int_{-h_i}^{t-h_i} \Phi(t - \tau) A_i \, d\tau
\]

for \(t > 0\), and assume that \(\Phi_f(0) = I\) and \(\Phi_f(t) = 0\) for \(t < 0\). Then

\[
x(t, z_0, 0) = \Phi_0(t)x(0) + \sum_{i=1}^{M} \int_{-h_i}^{t-h_i} \Phi_f(t - \tau - h_i) A_i x_0(\tau) \, d\tau
\]

is the free solution that depends only on the initial complete state \(z_0 = (x(0), x_0)\). Therefore the solution (2) takes the form

\[
x(t) = x(t, z_0, 0) + \int_0^t \Phi_f(t - \tau) B u(\tau) \, d\tau.
\]

Applying the methods presented by Manitius (1974) for integer-order systems with one delay in the state, we state the following theorem that formulates a new criterion for relative controllability of the system (1).

**Theorem 3.** The fractional system (1) is relatively controllable on the time interval \([0, T]\) if and only if the relation

\[
a^0 \Phi_f(T - t) B = 0
\]

for \(a \in \mathbb{R}^n\) and \(t \in [0, T]\) implies \(a = 0\).

**Proof.** Consider the linear mapping

\[
F: L^2([0, T], \mathbb{R}^m) \to \mathbb{R}^n
\]

given by

\[
F(u) = \int_0^T \Phi_f(T - \tau) B u(\tau) \, d\tau.
\]

Then \(F\) is a continuous linear operator from the Hilbert space \(L^2([0, T], \mathbb{R}^m)\) to the Hilbert space \(\mathbb{R}^n\). Let \(im F\) denote the range (image) of \(F\). Since the range of linear mapping is a subset of its codomain, we have

\[
im F \subset \mathbb{R}^n,
\]

and the orthogonal complement of the linear subspace \(im F\) satisfies the relation

\[
(im F)^\perp = \ker F,
\]

where \(F^\perp\) is the adjoint of \(F\) and \(\ker F\) denotes the null space (kernel) of \(F\).

Since the controllability condition becomes

\[
(im F)^\perp = \{0\},
\]

we obtain

\[
\ker F = 0.
\]

However, for any \(a \in \mathbb{R}^n\) and \(u \in L^2([0, T], \mathbb{R}^m)\), the scalar products in the given spaces are equal, that is,

\[
\langle a, F(u) \rangle_{\mathbb{R}^n} = \langle F(a), u \rangle_{L^2([0, T], \mathbb{R}^m)}.
\]

Therefore, we have

\[
\langle a, \int_0^T \Phi_f(T - \tau) B u(\tau) \, d\tau \rangle_{\mathbb{R}^n} = \int_0^T \langle a^0 \Phi_f(T - \tau) B \rangle u(\tau) \, d\tau.
\]

Thus \(\langle a, F(u) \rangle_{\mathbb{R}^n} = \langle a^0 \Phi_f(T - t) B \rangle u(t)\) for \(t \in [0, T]\). It follows that \(\ker F\) is a set of \(a \in \mathbb{R}^n\) such that \(a^0 \Phi_f(T - t) B = 0\) almost everywhere in \([0, T]\). But we have obtained above that \(\ker F\) consists of zero only (\(\ker F = 0\)), which proves the theorem.

In order to formulate the next criterion, for \(t \in [0, \infty)\), we define recursively the following matrices:

\[
Q_k(t) = \sum_{i=0}^{M} A_i Q_{k-1}(t - h_i), \quad k = 1, 2, \ldots.
\]

and the set

\[
\hat{Q}_n(T) = \{Q_0(t), Q_1(t), \ldots, Q_{n-1}(t), \quad t \in [0, T]\}
\]

for \(t = h_1, 2h_1, 3h_1, \ldots \); \(i = 0, 1, 2, \ldots, M\).

Let rank \(\hat{Q}_n(T)\) mean the rank of the block matrix composed of all matrices from the set \(\hat{Q}_n(T)\).

The theorem below is the necessary and sufficient condition for relative controllability of the system (1) that is based on the matrices \(A_0, \ldots, A_M\) and \(B\). This, easy
to use, algebraic criterion is similar to the commonly known Kalman rank condition. The criterion is a generalization of the controllability condition formulated for integer-order systems with delays. Details can be found in the work of Klamka (1991).

**Theorem 4.** The fractional system (1) is relatively controllable on the time interval \([0, T]\) if and only if
\[
\text{rank} \hat{Q}_n(T) = n.
\]

**Remark 1.** It is worth noticing that for \(T \leq h_1\) we obtain
\[
\hat{Q}_n(T) = \{ B, A_0 B, A_0^2 B, \ldots, A_0^{n-1} B \},
\]
and the condition \(\text{rank} \hat{Q}_n(T) = n\) is reduced to well-known controllability criteria for fractional systems without delays (see, e.g., Monje et al., 2010)
\[
\text{rank} [B A_0 B A_0^2 B \ldots A_0^{n-1} B] = n.
\]

Now we impose constraints on control values. Let \(U \subset \mathbb{R}^m\) be a nonempty set and \(S \subset \mathbb{R}^n\) have the form
\[
S = \{ x \in \mathbb{R}^n : Lx = c \}, \tag{6}
\]
where \(L\) is a \((p \times n)\)-matrix and \(c \in \mathbb{R}^p\) is a given vector. If \(L = I_n\) (identity matrix) and \(c = 0\), we obtain \(S = \{ 0 \} \).

In much the same way as for integer-order dynamical systems (Klamka, 1991), we can formulate a definition of the attainable set for the dynamical system (1).

**Definition 5.** The set
\[
K([0, t], z_0) = \left\{ x \in \mathbb{R}^n : x(t) = x(t, z_0, 0) \right\}
\]
is called an attainable set from the initial complete state \(z_0 = (x(0), x_0)\) for the time-delay fractional system (1). \(K([0, t], z_0)\) is also called the set of reachable states.

The next theorem gives a new criterion for relative U-controllability of the fractional system (1) when \(U\) is a convex and compact subset of \(\mathbb{R}^m\) containing the origin.

**Theorem 5.** Let \(U \subset \mathbb{R}^m\) be a convex and compact set containing the origin in its interior and let the system (1) be of commensurate order. If
\[
\text{rank} \hat{Q}_n(T) = n,
\]
then the fractional system (1) is relatively null-controllable on \([0, T]\).
for some finite $T \in (0, \infty)$, where $N(0)$ is a sufficiently small neighborhood of $0 \in \mathbb{R}^n$. Then, the instantaneous state $x(T, z_0, 0)$ can be driven to $0 \in \mathbb{R}^n$ in a finite time, so that the fractional system (1) is relatively null $U$-controllable. ■

In order to formulate the next criteria of relative $U$-controllability of the fractional system (1), we introduce a scalar function $J : \mathbb{R}^n \times \mathbb{R} \times \mathbb{R}^p \rightarrow \mathbb{R}$, related to the attainable set $K([0, T], z_0)$ of the system (1) and defined as follows:

$$J(z_0, T, a) = a' Lx(t, z_0, 0) + \int_0^T \sup \left\{ a' L \Phi_f(t - \tau) Bu(\tau), \right\} d\tau - a' c,$$  

(7)

where $a \in \mathbb{R}^p$ is any vector and $a'$ is its transpose. The function $J$ is called a supporting function of the attainable set. Applications of the supporting function for integer-order systems can be found in the works of Klamka (2012). Let $\bar{K}(t)$ be the solution corresponding to $\bar{u}(t)$. Then in $[0, T]$ we have

$$\bar{x}(t) = x(t, z_0, 0) + \int_0^T \Phi_f(t - \tau) \bar{u}(\tau) d\tau = \lim_{k_i \to \infty} x_{k_i}(t),$$

and

$$\lim_{k_i \to \infty} x_{k_i}(T) = \bar{x}(T) \in K([0, T], z_0),$$

which proves the compactness.

It follows that the set $K([0, T], z_0)$ of the form

$$K([0, T], z_0) = \{ y \in \mathbb{R}^p : y = Lx, x \in K([0, T], z_0) \}$$

is also convex and compact. An initial complete state $z_0$ can be driven to the set $S$ in time $T > 0$ if and only if the vector $c$ and the set $K([0, T], z_0)$ cannot be strictly separated by a hyperplane, that is, if

$$a' c \leq \sup \{ a' \bar{x} : \bar{x} \in \bar{K}([0, T], z_0) \}$$

for all vectors $a \in \mathbb{R}^p$.

Since $\bar{x} \in \bar{K}([0, T], z_0)$, we have $\bar{x} = Lx$ for any $x \in K([0, T], z_0)$, that is,

$$\bar{x} = L \left( x(t, z_0, 0) + \int_0^t \Phi_f(t - \tau) Bu(\tau) d\tau \right).$$

Therefore the above inequality can be equivalently written in the following form:

$$a' Lx(t, z_0, 0) + \int_0^T \sup \left\{ a' L \Phi_f(t - \tau) Bu(\tau), \right\} d\tau - a' c \geq 0.$$

Interchanging integration and the supremum operation, we conclude that $c \in \bar{K}_U([0, T], z_0)$ if and only if $J(z_0, T, a) \geq 0$ for all $a \in \mathbb{R}^p$. Moreover, we can show that

$$k J(z_0, T, a) = J(z_0, T, ka) \quad \text{for every } k \geq 0.$$

Therefore, for vectors $a \in E$, the proof is complete. ■

Corollary 1. Let $U \subset \mathbb{R}^m$ be a compact set and $E \subset \mathbb{R}^n$ be any set containing the origin as an interior point. Then the fractional dynamical system (1) is relatively null $U$-controllable $z_0 \in \mathbb{R}^n \times L^2([-h_M, 0], \mathbb{R}^n)$ if and only if for some $T \in [0, \infty)$

$$J(z_0, T, a) \geq 0 \quad \text{for every } a \in E.$$
The corollary is an immediate consequence of Theorem 4 for $S = \{0\}$, i.e., for $L = I_n$ and $c = 0$. $E$ is then a subset of $\mathbb{R}^n$.

**Theorem 7.** If $U$ is a compact set containing the origin, then the fractional system (1) is relatively null $U$-controllable if and only if the equality

$$
\int_0^{+\infty} \sup \{ a'\Phi_f(T - \tau)Bu(\tau) : u \in L^2_{\text{loc}}([0, T], U) \} \, d\tau = +\infty \tag{8}
$$

holds for every nonzero vector $a \in \mathbb{R}^n$ and $T > 0$.

**Proof.** Let us prove the necessary condition by contradiction. Assume that the retarded fractional system (1) is relatively null $U$-controllable and the condition (8) is not satisfied. Then there exists a constant $k$, $0 < k < +\infty$, and a nonzero vector $\tilde{a} \in \mathbb{R}^n$ such that, for all $T > 0$,

$$
\int_0^T \sup \{ \tilde{a}'\Phi_f(T - \tau)Bu(\tau) : u \in L^2([0, T], U) \} \, d\tau < k.
$$

For $S = \{0\}$ the supporting function takes the form

$$J(z_0, T, a) = a'x(t, z_0, 0) + \int_0^T \sup \{ a'\Phi_f(T - \tau)Bu(\tau) : u \in L^2([0, T], U) \} \, d\tau.
$$

We put

$$x(t, z_0, 0) = -2k\tilde{a}
$$

for some initial conditions $\tilde{z}_0$.

For $\tilde{a} \neq 0$, also $x(t, \tilde{z}_0, 0) \neq 0$ and we have

$$J(\tilde{z}_0, T, \tilde{a}) = \tilde{a}'x(t, \tilde{z}_0, 0) + \int_0^T \sup \{ \tilde{a}'\Phi_f(T - \tau)Bu(\tau) : u \in L^2([0, T], U) \} \, d\tau = -2k \\sup \{ \tilde{a}'\Phi_f(T - \tau)Bu(\tau) : u \in L^2([0, T], U) \} \, d\tau < -k.
$$

Hence, for all $T \geq 0$, we obtain $J(\tilde{z}_0, T, \tilde{a}) < 0$, which implies that the retarded system (1) is not relatively null $U$-controllable from the initial complete conditions $\tilde{z}_0$ (see Corollary 1). This contradicts the assumption that (1) is relative null $U$-controllable. In this way, the necessary condition is proved.

The sufficient condition will be also proved by contradiction. Assume that (8) holds and the retarded fractional system (1) is not relatively null $U$-controllable. Thus these are initial conditions $\tilde{z}_0$ from which the system cannot be driven into zero. Therefore, for some $T > 0$, there exists $\tilde{a} \neq 0$ such that the following inequality is satisfied:

$$\tilde{a}'x(t, \tilde{z}_0, 0) + \int_0^T \sup \{ \tilde{a}'\Phi_f(T - \tau)Bu(\tau) : u \in L^2([0, T], U) \} \, d\tau < 0.
$$

Applying the Schwarz inequality (see Rolewicz, 1987) we have

$$\int_0^T \sup \{ \tilde{a}'\Phi_f(T - \tau)Bu(\tau) : u \in L^2([0, T], U) \} \, d\tau \leq ||\tilde{a}|| ||x(t, \tilde{z}_0, 0)||,
$$

which contradicts the assumption that (8) is true. This completes the proof.

The next theorem follows from Theorem 4 and provides the controllability criterion for nonnegative constraints.

**Corollary 2.** If $U = \mathbb{R}^m_{+} \cup \{0\}$, then the fractional system (1) is relatively null $U$-controllable if and only if the equality

$$\int_0^{+\infty} \sup \{ a'\Phi_f(T - \tau)Bu(\tau) : u \in L^2_{\text{loc}}([0, \infty), U) \} \, d\tau = +\infty
$$

holds for every nonzero vector $a \in \mathbb{R}^n$ and $T > 0$.

**Proof.** The corollary follows immediately from Theorem 4 since the first orthant in the $m$-dimensional Euclidean space $\mathbb{R}^m$ includes a compact subset containing $0 \in \mathbb{R}^m$.

Nonnegative or positive controls are especially important in practical applications. Some controllability criteria for continuous-time linear fractional systems with positive controls are presented by Kaczorek (2014a; 2014b). However they concern positive systems. Here we consider arbitrary systems, which means that other parameters of the systems do not have to be positive. Continuous-time positive linear systems are also discussed by Zhao et al. (2013; 2014) in the case of one time-delay and without delays, respectively.

5. **Examples**

The numerical examples below illustrate the theoretical results presented in the paper.
Example 1. Consider a linear fractional system described by the following fractional differential equation:

\[ C \frac{d^\tau}{dt^\tau} x(t) = A_0 x(t) + A_1 x(t - 1) + A_2 x(t - 2) + B u(t), \quad (9) \]

for \( t \in [0, 3] \) with the initial conditions \( x(0) = \{x(0), z_0\}, \) where

\[
A_0 = \begin{bmatrix}
0 & 2 \\
0 & 0
\end{bmatrix}, \\
A_1 = \begin{bmatrix}
0 & 0 \\
0 & -2
\end{bmatrix}, \\
A_2 = \begin{bmatrix}
0 & 1 \\
-1 & -3
\end{bmatrix}, \\
B = \begin{bmatrix}
1 \\
-1
\end{bmatrix}.
\]

We have \( n = 2, M = 3, h_0 = 0, h_1 = 1, h_2 = 3. \) In order to verify whether the system (9) is relatively controllable on \([0, 3]\), we apply Theorem 6. We will show that the matrix \( W(0, 3) \) is nonsingular. This means that the matrix has to be full rank, i.e., \( \text{rank} W(0, 3) = 2. \) Thus we have to show that

\[
\text{rank} \int_0^3 \Phi(3 - \tau)BB'\Phi'(3 - \tau) d\tau = 2.
\]

Using the Cayley–Hamilton method (see Monje et al., 2010) we calculate

\[
\Phi(t) = t^{-\frac{1}{2}} \sum_{k=0}^{1} \frac{A_k^0}{\Gamma\left(\frac{1}{2}k + 1\right)} t^{\frac{1}{2}k} + t^{-\frac{1}{2}} \begin{bmatrix}
0 & 2 \\
0 & 0
\end{bmatrix} t^1,
\]

and

\[
\Phi'(t) = \sum_{k=0}^{1} \frac{(A_k^0)^k}{\Gamma\left(\frac{1}{2}k + 1\right)} t^{\frac{1}{2}k} + \begin{bmatrix}
0 & 2 \\
0 & 0
\end{bmatrix} t^1.
\]

It follows that

\[
W(0, 3) = \int_0^3 \begin{bmatrix}
\frac{1}{\sqrt{\pi}}(3 - \tau)^{-\frac{1}{2}} & 2 \\
0 & \frac{1}{\sqrt{\pi}}(3 - \tau)^{-\frac{1}{2}}
\end{bmatrix}
\times \begin{bmatrix}
1 & -1 \\
-1 & 1
\end{bmatrix}
\begin{bmatrix}
\frac{1}{\sqrt{\pi}}(3 - \tau)^{-\frac{1}{2}} & 0 \\
0 & \frac{1}{\sqrt{\pi}}(3 - \tau)^{-\frac{1}{2}}
\end{bmatrix} d\tau,
\]

since

\[
BB' = \begin{bmatrix}
1 & -1 \\
-1 & 1
\end{bmatrix}.
\]

Multiplying the matrices under the integral and integrating each element of the result matrix, we obtain

\[
W(0, 3) = \begin{bmatrix}
\frac{2\sqrt{\pi}}{\pi} - \frac{12\sqrt{\pi}}{\sqrt{3} - 3\sqrt{3}} + 4\sqrt{3} & \frac{4\sqrt{2}}{\sqrt{3} - 3\sqrt{3}} \\
\frac{2\sqrt{\pi}}{\sqrt{3} - 3\sqrt{3}} & \frac{2\sqrt{2}}{\sqrt{3} - 3\sqrt{3}}
\end{bmatrix}.
\]

We see that \( \text{rank} W(0, 3) = 2 \), which implies the relative controllability of the fractional system (9) on \([0, 3]\).

Example 2. Let a fractional system with two delays in the state be described by the equation

\[ C \frac{d^\tau}{dt^\tau} x(t) = A_0 x(t) + A_1 x(t - 1) + A_2 x(t - 2) + B u(t), \]

on any interval \([0, T]\), for

\[
A_0 = \begin{bmatrix}
0 & 1 \\
-1 & -2
\end{bmatrix}, \quad A_1 = \begin{bmatrix}
0 & 0 \\
0 & -2
\end{bmatrix}, \quad A_2 = \begin{bmatrix}
0 & 0 \\
3 & 0
\end{bmatrix}, \quad B = \begin{bmatrix}
0 \\
-1
\end{bmatrix}.
\]

We apply Theorem 6 to verify relative controllability of the system (10) on the interval \([0, T]\). Since \( n = 2 \), we have to show that \( \text{rank} \tilde{Q}_2(T) = 2 \) for each \( T > 0 \).

Let us find all matrices \( Q_k(t) \) belonging to the set \( \tilde{Q}_2(T) \), defined by the formula (5). We have

\[
Q_0(0) = B = \begin{bmatrix}
0 \\
-1
\end{bmatrix},
\quad Q_0(t) = 0
\]

for \( t \neq 0 \). Next, since

\[
Q_1(t) = \sum_{i=0}^{2} A_i Q_0(t - h_i)
\]

for \( t = h_i, 2h_i, 3h_i, \ldots \) and \( i = 0, 1, 2 \), we calculate

\[
Q_1(0) = A_0 B = \begin{bmatrix}
0 & 1 \\
-1 & -2
\end{bmatrix} \begin{bmatrix}
0 \\
-1
\end{bmatrix} = \begin{bmatrix}
-1 & 2
\end{bmatrix},
\quad Q_1(h_i) = A_1 Q_0(0) = \begin{bmatrix}
0 & 0 \\
0 & -2
\end{bmatrix} \begin{bmatrix}
0 & 0 \\
0 & -1
\end{bmatrix} = \begin{bmatrix}
0 & 2
\end{bmatrix},
\quad Q_1(h_2) = A_2 Q_0(0) = \begin{bmatrix}
0 & 0 \\
3 & 0
\end{bmatrix} \begin{bmatrix}
0 & 0 \\
0 & -1
\end{bmatrix} = \begin{bmatrix}
0 & 0
\end{bmatrix},
\quad Q_1(2h_i) = A_2 Q_0(0) = Q_1(h_2),
\]

since \( 2h_1 = h_2 \). Other matrices \( Q_1(jh_i), j = 2, 3, \ldots \), are equal to zero. Therefore

\[
\tilde{Q}_2(T) = \{Q_0(0), Q_1(0), Q_1(h_1), Q_1(h_2)\},
\]

and finally

\[
\text{rank} \tilde{Q}_2(T) = \text{rank} \begin{bmatrix}
0 & -1 & 0 & 0 \\
-1 & 2 & 2 & -3
\end{bmatrix} = 2.
\]

Based on Theorem 6 we conclude that the delayed fractional system (10) is relatively controllable on any interval \([0, T]\).
6. Concluding remarks

Relative controllability and relative constrained controllability of linear fractional systems with delays in the state were discussed in the paper. Constraints imposed on the delay values were considered. The formula for a solution of the discussed systems was derived with the use of the Laplace transform (Theorem 1). Definitions of relative controllability for unconstrained as well as for constrained controls were formulated. The contribution of the paper consists of several new necessary and sufficient conditions for relative controllability (Theorems 3, 4, 5, 6 and relative $U$-controllability (Theorem 7, Corollary 3) for time-delay fractional systems described by Eqn. (1), which were established and proved in detail. Numerical examples were presented to illustrate how to verify relative controllability of the discussed systems with the use of the established criteria. The presented theoretical results can be extended to semilinear fractional systems with retarded controls.

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